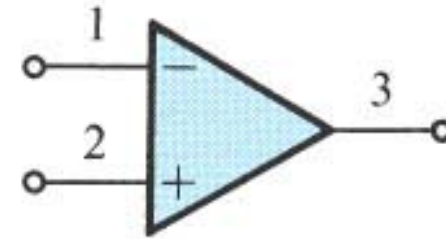
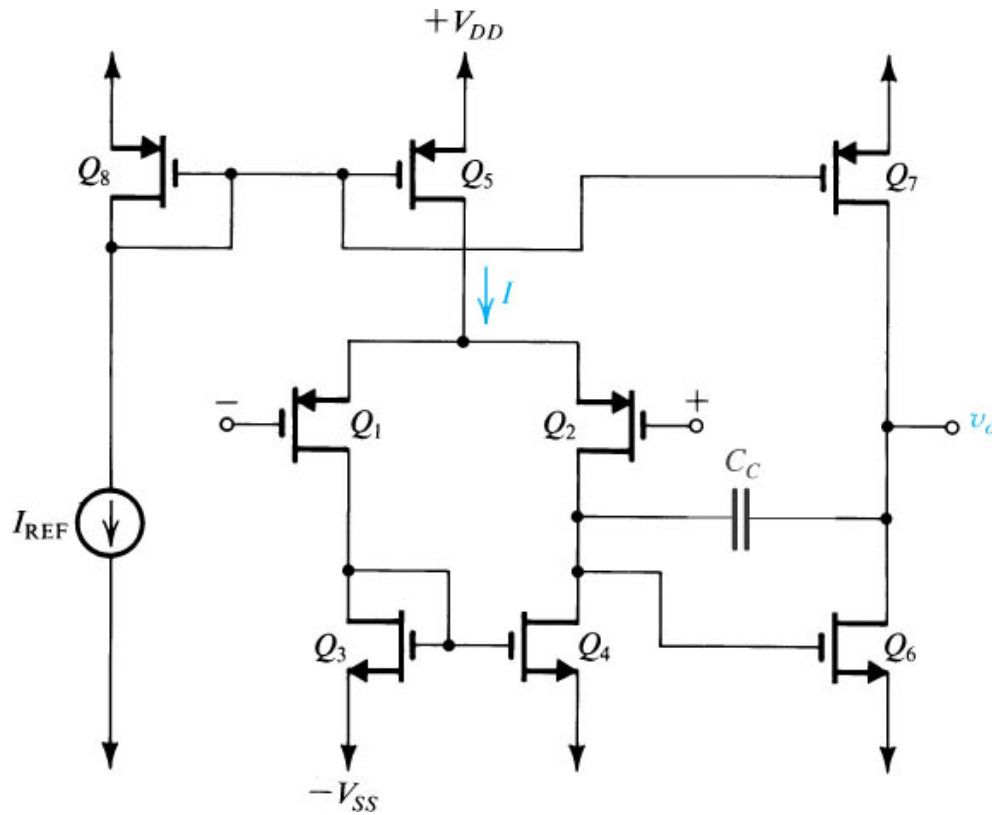


# Lect. 16: Operational Amplifier

(Chap. 8 in Razavi)

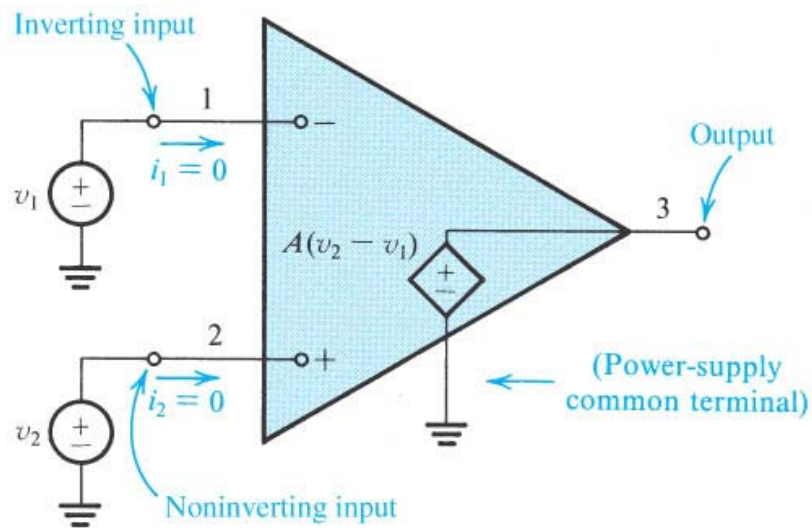


Operational Amplifier (Op Amp)

What is the application for this?

# Lect. 16: Operational Amplifier

Characteristics of an *ideal* op amp



- Ideal amplifier for input voltage difference

$R_{in}$ : Infinite

$A_v$ : Infinite

$R_{out}$ : 0 for voltage output

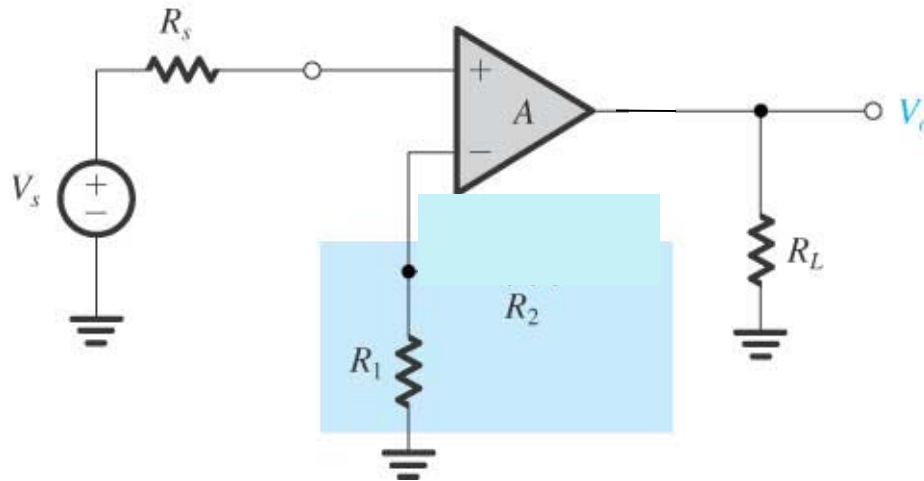
infinite for current output

→ OTA

Zero common-mode gain  
(or infinite common-mode rejection)

# Lect. 16: Operational Amplifier

Op amp is very often used with feedback



Remember  $A_f = \frac{x_o}{x_s} = \frac{A}{1 + A\beta} \sim \frac{1}{\beta}$

$$\beta = R_1 / (R_1 + R_2)$$

$$V_o = A_v(V_s - 0)$$

→ +/- supply voltage

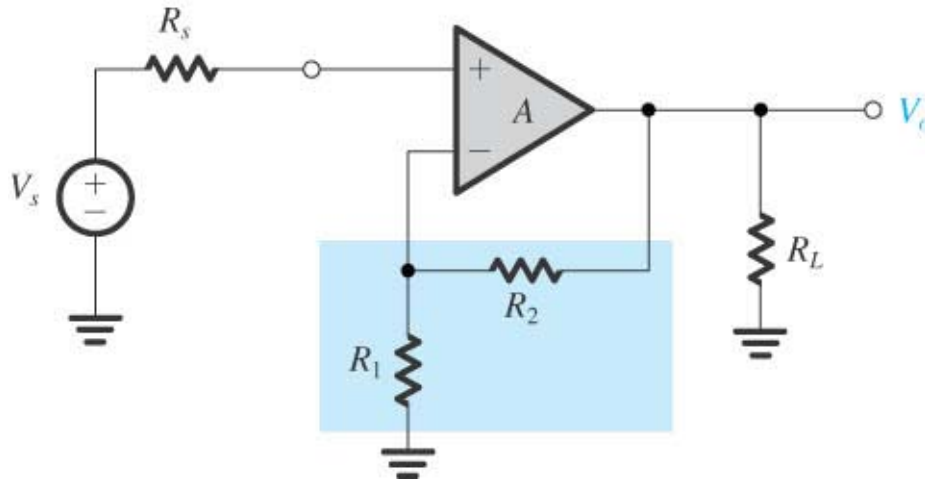
With feedback

$$V_o = A_v \left( V_s - V_o \cdot \frac{R_1}{R_1 + R_2} \right)$$

$$V_o \left( 1 + \frac{A_v R_1}{R_1 + R_2} \right) = A_v V_s$$

$$\therefore \frac{V_o}{V_s} = \frac{A_v}{1 + \frac{A_v R_1}{R_1 + R_2}} \sim \frac{R_1 + R_2}{R_1}$$

# Lect. 16: Operational Amplifier



$$\frac{V_o}{V_s} = \frac{R_1 + R_2}{R_1}$$

The same result can be obtained by assuming  $V^+ = V^-$  (Virtual Short)

$$V_s = V_o \frac{R_1}{R_1 + R_2}$$

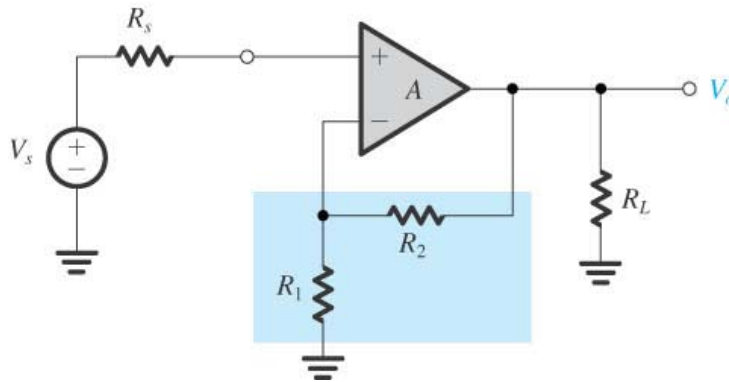
→ Feedback tries to maintain  $V^+ = V^-$

$$\therefore \frac{V_o}{V_s} = \frac{R_1 + R_2}{R_1}$$

Use virtual short condition for Op-Amp analysis!

# Lect. 16: Operational Amplifier

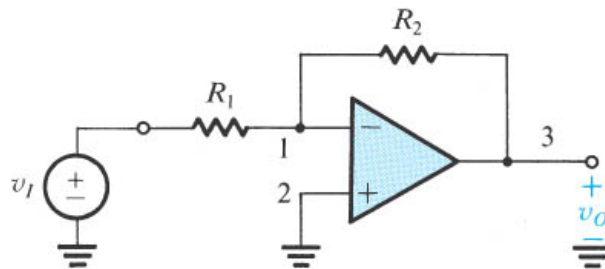
What good is it?



$$\frac{V_o}{V_s} = \frac{R_1 + R_2}{R_1} \quad \text{Voltage amplifier}$$

- Infinite input resistance
- Same gain regardless of  $R_L$
- Gain is stable and can be easily changed

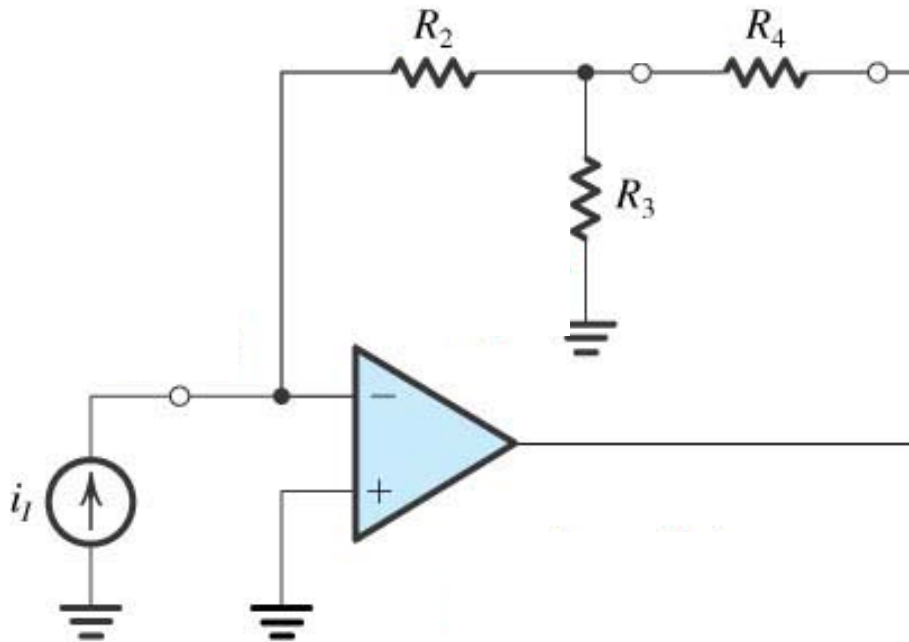
Voltage amplifier with negative gain?



$$V_o = -\frac{R_2}{R_1} V_I$$

# Lect. 16: Operational Amplifier

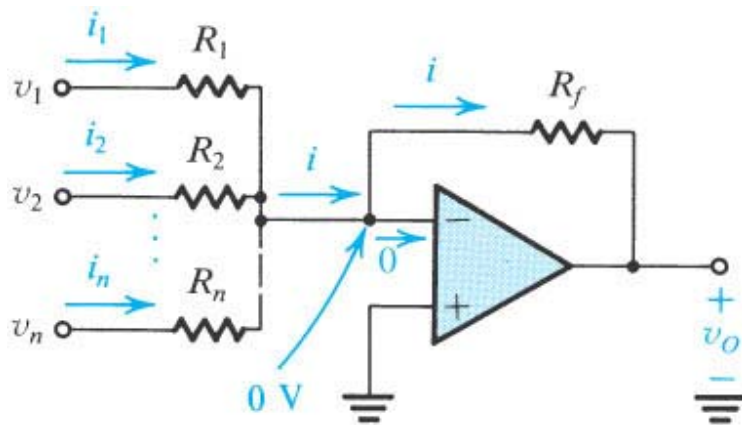
Current amplifier



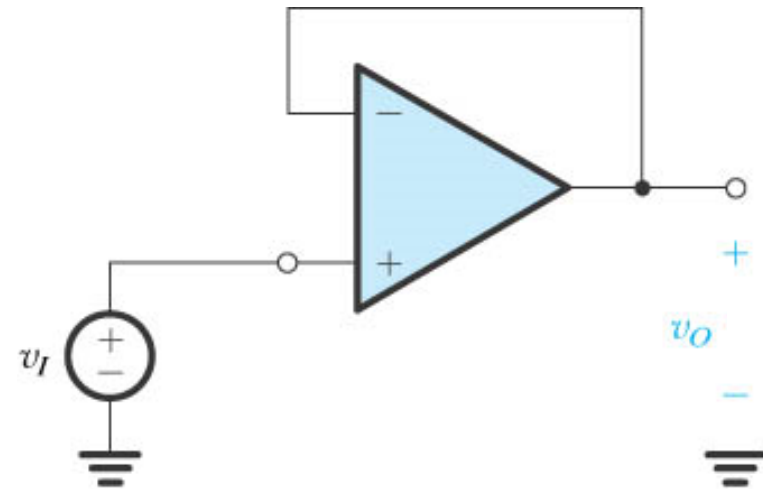
$$A_i = \frac{i_4}{i_i} = \left(1 + \frac{R_2}{R_3}\right)$$

No loading effect!

# Lect. 16: Operational Amplifier



weighted summer (adder)

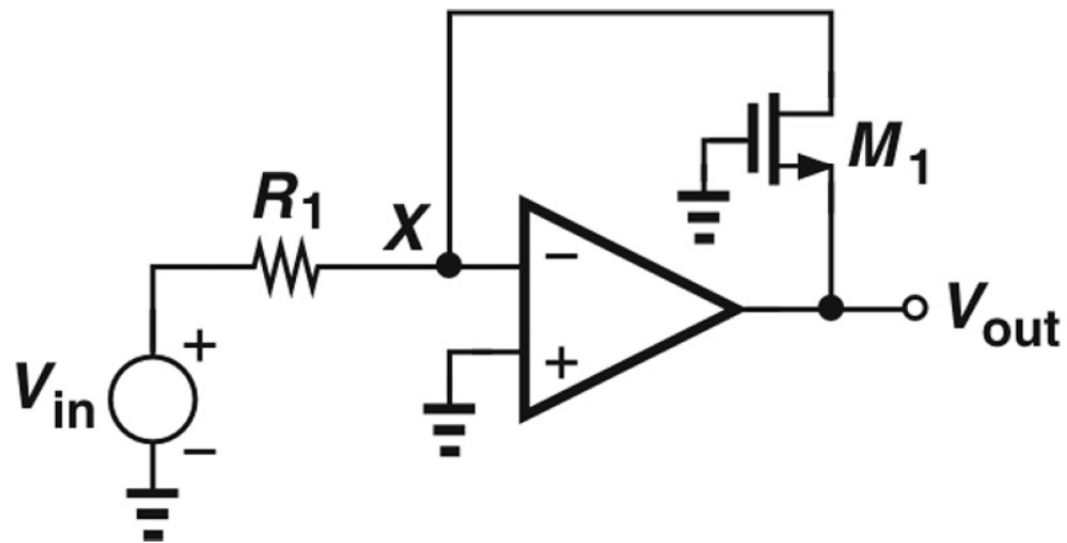


(a)

voltage buffer

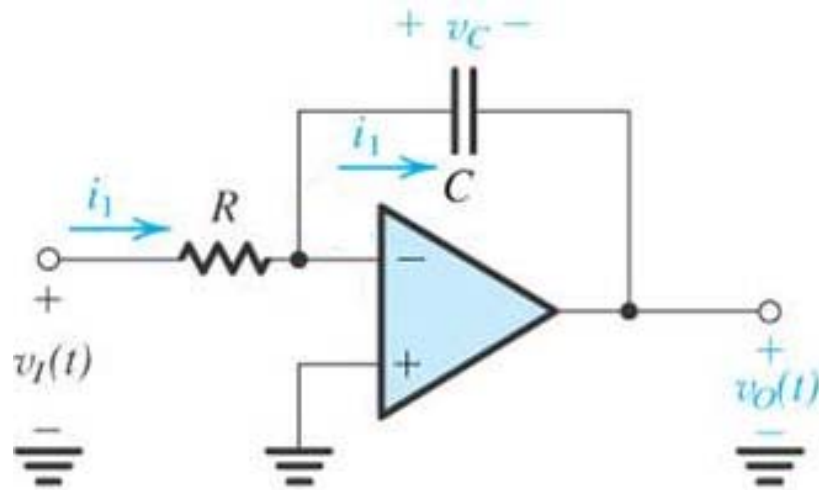
# Lect. 16: Operational Amplifier

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# Lect. 16: Operational Amplifier



$$C \frac{dv_o(t)}{dt} = -\frac{v_i(t)}{R}$$

$$v_o(t) - v_o(t=0) = -\frac{1}{RC} \int_0^t v_i(t) dt$$

→ Integrator

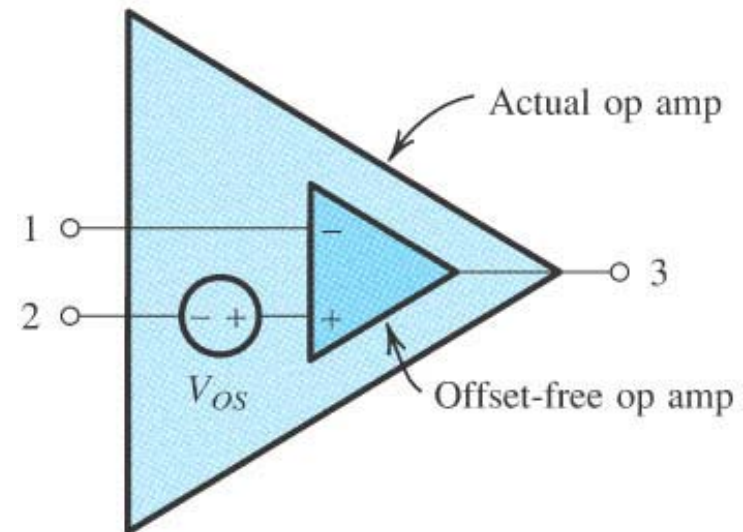
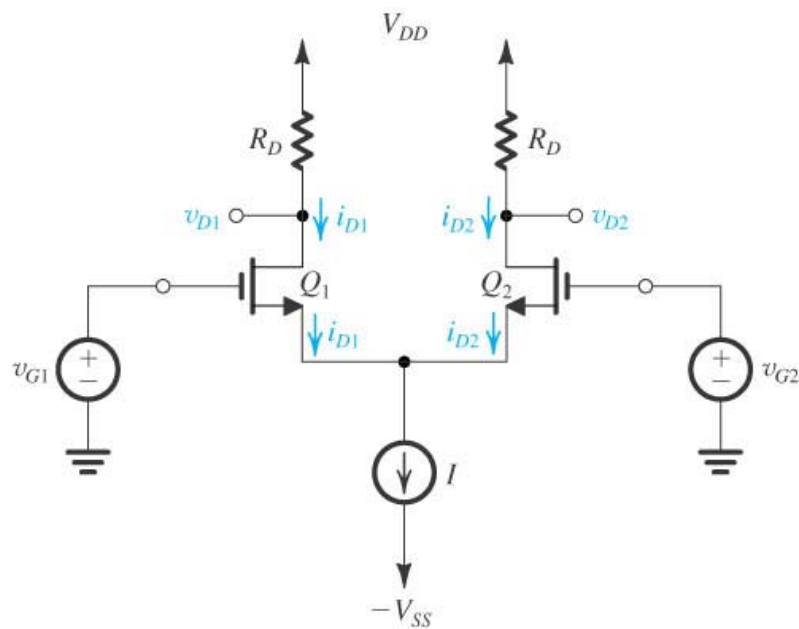
In s-domain?

# Lect. 16: Operational Amplifier

In reality, op amps have input offset

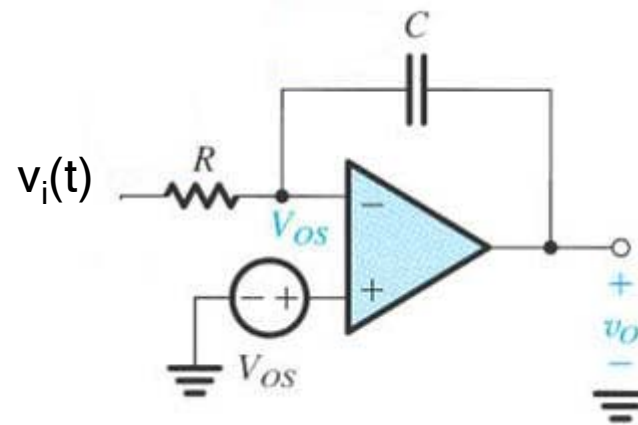
Ideally,  $v_o = 0$  if  $v^+ = v^-$  But, often,  $v_o \neq 0$  when  $v^+, v^-$  have offset voltage  $V_{os}$

Caused by mismatched in differential pair

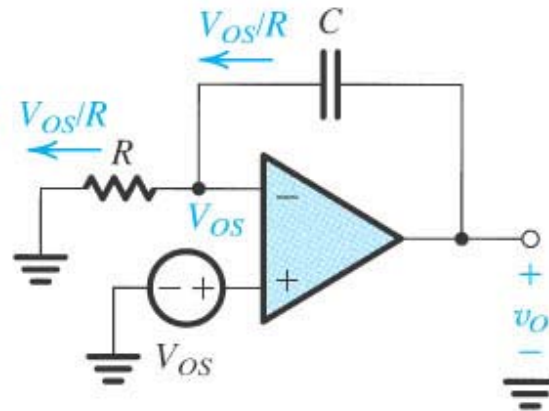


# Lect. 16: Operational Amplifier

What happens to Op Amp integrator?



Consider zero input



$$\begin{aligned} v_o &= V_{os} + \frac{1}{C} \int_0^t \frac{V_{os}}{R} dt \\ &= V_{os} + \frac{V_{os}t}{CR} \end{aligned}$$

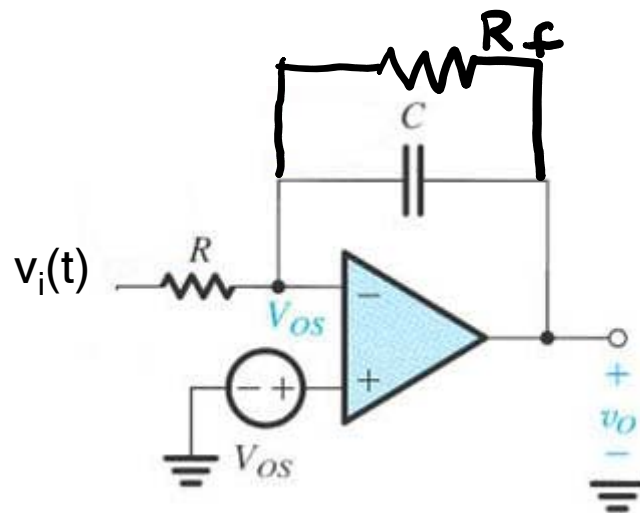
It does not work as an integrator!

How can you solve this problem?

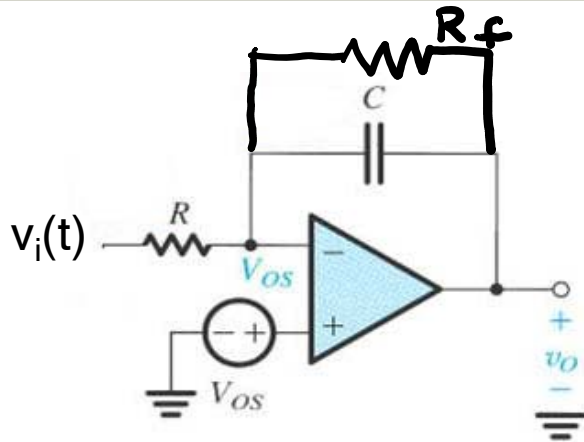
# Lect. 16: Operational Amplifier

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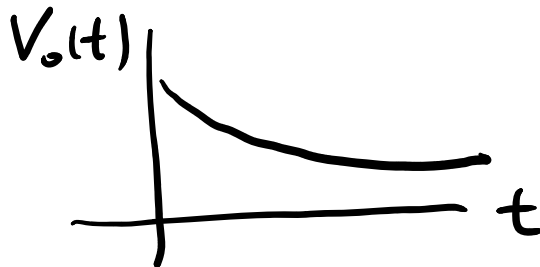
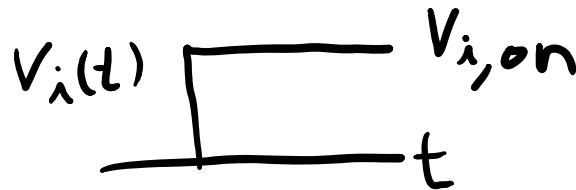
Solution: Lossy Integrator



# Lect. 16: Operational Amplifier



Is this an integrator?



Yes, if  $t \ll R_f C$

$\rightarrow f \gg 1/(R_f C)$