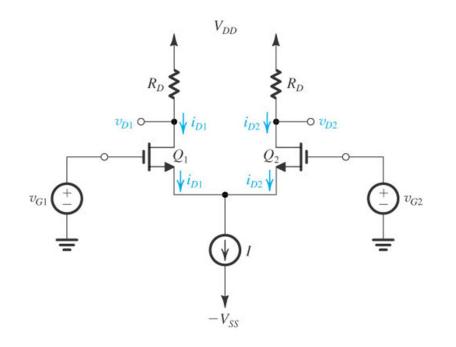
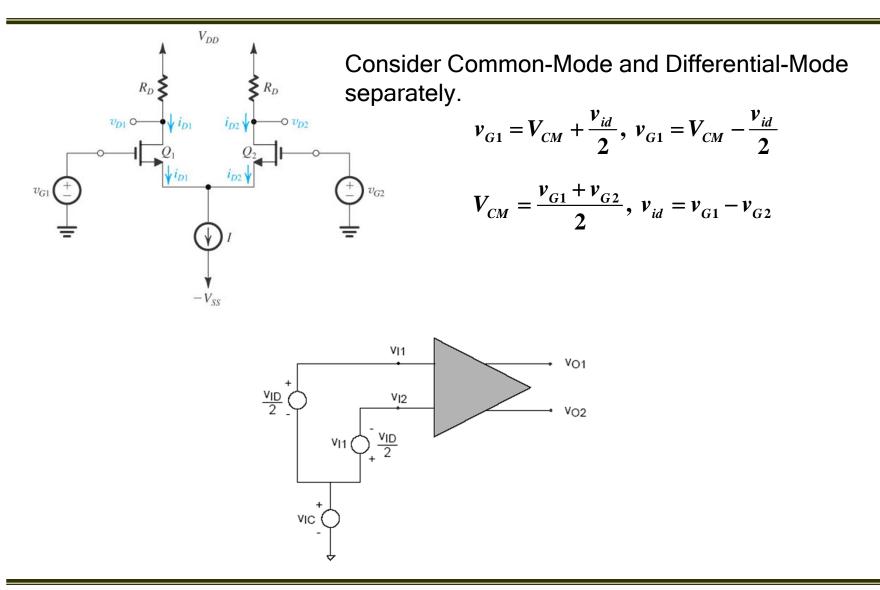
Differential Amplifier



If
$$v_{G1} = v_{G2}$$
, $v_o = v_{D1} - v_{D2} = 0$
If $v_{G1} > v_{G2}$, $v_o < 0$
If $v_{G1} < v_{G2}$, $v_o > 0$

Non-zero output only for input difference → Differential amplifier







Common-Mode: $v_{G1} = v_{G2} = v_{CM}$ V_{DD} R_D $I/2 \bigvee v_{D2} = V_{DD} - \frac{I}{2}R_D$ $v_{D1} = V_{DD} - \frac{I}{2}R_D \circ \psi I/2$ UCM $v_S = v_{CM} - V_{GS}$ $-V_{SS}$

 $v_{D1} - v_{D2} = 0$ (No CM output)

v_{CM, max}?

$$v_{DS} \ge v_{GS} - V_{TH}$$

$$(V_{DD} - \frac{I}{2}R_D) - (v_{CM} - v_{GS}) \ge v_{GS} - V_{TH}$$

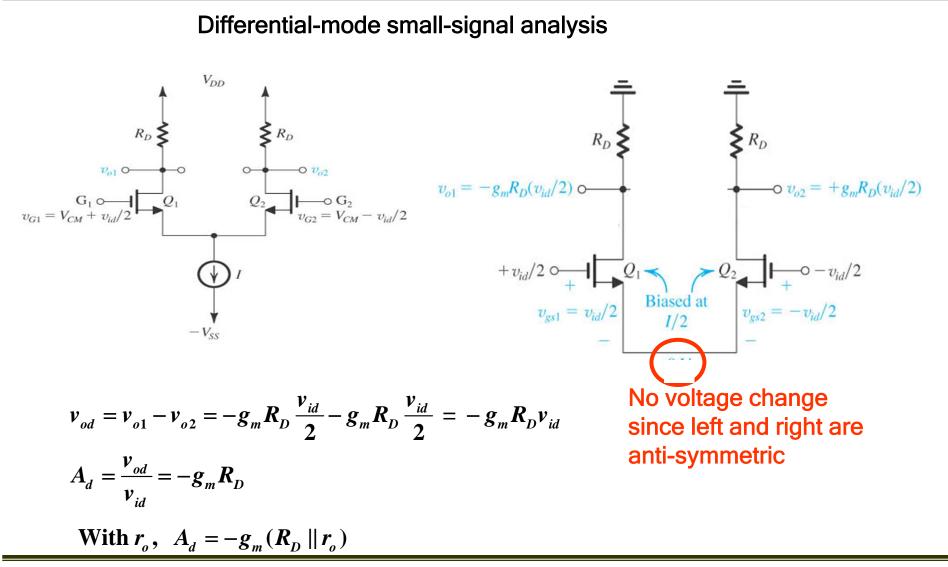
$$V_{DD} - \frac{I}{2}R_D + V_{TH} \ge v_{CM}$$

$$\therefore v_{CM,\max} = V_{TH} + V_{DD} - \frac{I}{2}R_D$$

 $v_{CM, min}$?

➔ Input common-mode range





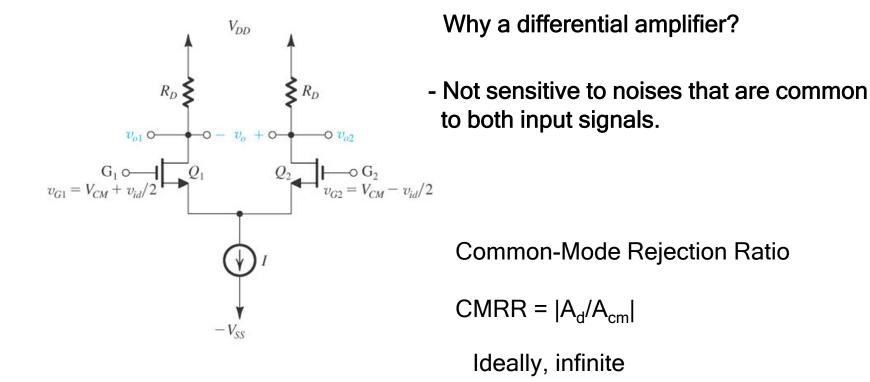


Differential-mode small-signal half-circuit → Consider only half circuit



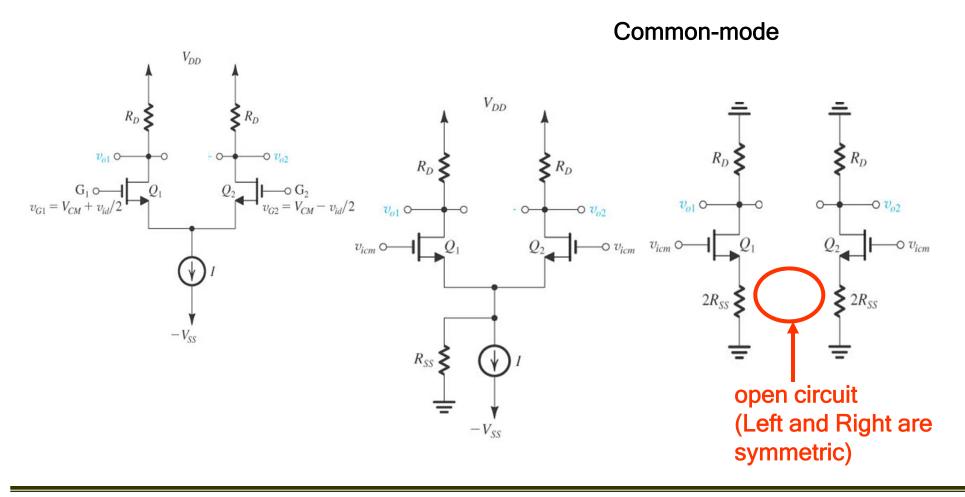
Differential pair acts as CS amplifier for input difference!







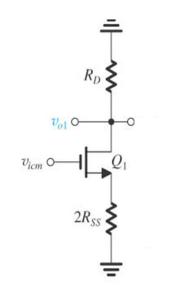
What if the current source is NOT ideal?





Common-Source with source resistance

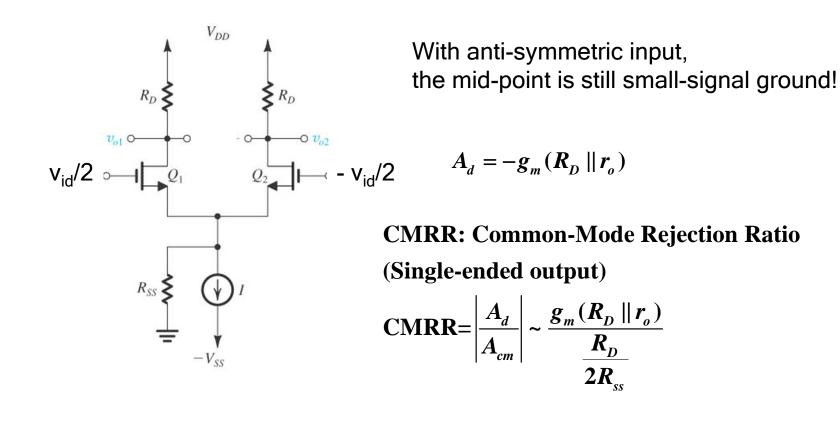
Common-mode half circuit



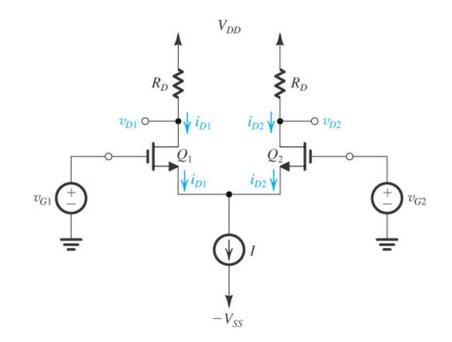
 $\frac{v_{o1}}{v_{icm}} = ? \qquad v_{o1} = -g_m v_{gs} R_D$ $v_{gs} = v_{icm} - g_m v_{gs} (2R_{SS})$ $v_{gs} = \frac{v_{icm}}{1 + 2g_m R_{SS}}$ $\frac{v_{o1}}{v_{icm}} = -\frac{g_m R_D}{1 + 2g_m R_{SS}} \sim -\frac{R_D}{2R_{SS}}$ $\Rightarrow \text{ common-mode gain due to } R_{ss} \text{ for single-ended output}$ Differential output ($v_0 = v_{01} - v_{02}$), $v_0 = 0$



What if the current source is NOT ideal? Differential-Mode







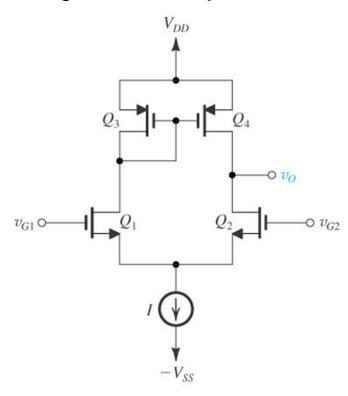
MOS differential pair is based on the symmetry

Anything that breaks the symmetry affects the circuit performance (CMRR, input offset)

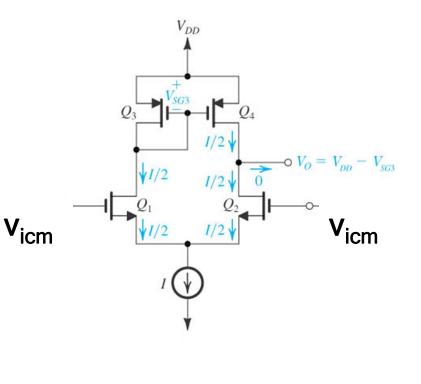
Device mismatches cause non-ideal performance \rightarrow Eliminate R_D



Active-loaded MOS differential pair for single-ended output

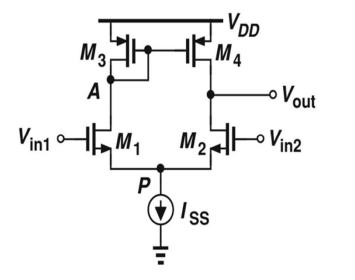




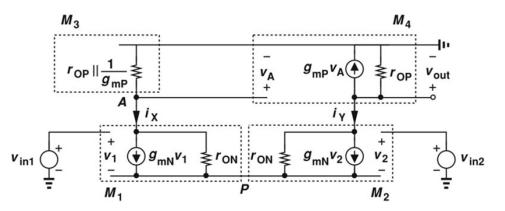




Differential mode small-signal analysis



Small-signal circuit



With complicated analysis (Razavi 10.6.2)

$$\frac{v_{out}}{v_{in1} - v_{in2}} = g_{mN} r_{ON} \frac{r_{OP} [1 + g_{mP} (\frac{1}{g_{mP}} || r_{OP})]}{2r_{ON} + 2r_{OP}}$$
$$\approx g_{mN} (r_{ON} || r_{OP})$$

Factor of 2 difference!

But the given circuit is asymmetric

→ Half circuit analysis is not possible

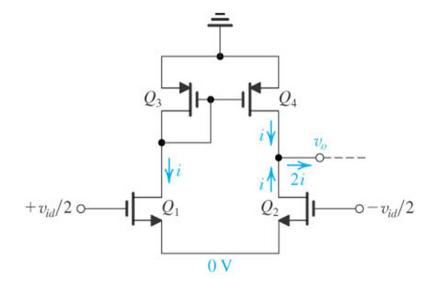
Electronic Circuits 2 (15/1)



Applying half-circuit analysis,

 $\frac{v_{out}}{v_{...} - v_{...}} = \frac{g_{mN}(r_{ON} || r_{OP})}{2}$

 $v_{in1} - v_{in2}$



The current mirror action is not considered with half-circuit analysis

Current mirror doubles the transconductance

→ Twice voltage gain!

